

Research Article A Novel 5G LTE Antenna Design for Portable Devices

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This paper presents the low profile, planar, and small-size antenna design for WWAN, LTE, and 5G (5th generation wireless systems) for use in portable communication equipment. The antenna occupies only $65 \times 13 \times 0.4$ mm³, and the antenna is combined with a 200×260 mm² copper plate to simulated system ground plane. In the low band, a direct-fed right-side arm and a coupled-fed arm implemented can excite a $1/4 \lambda$ fundamental resonant mode at 0.85 and 0.76 GHz to cover 0.698–0.96 GHz and upper $3/4 \lambda$ and $5/4 \lambda$ resonant modes are controlled by L-shaped element at 2.34, 2.69, 3.4, and 4.0 GHz to cover 1.71–2.69 GHz and 3.2-4.2 GHz. The direct-fed left-side arm produced $1/4 \lambda$ to cover 5.15-5.85 GHz. In far-field measured, peak gain and efficiency in low, middle, and high bands are 0.43–5.67 dBi and 55–86%. Finally, experiments demonstrate that the present antenna exhibits a good performance for portable devices.

1. Introduction

The evolution of wireless system technology has enabled not only the miniaturization but also the multiband applicability of antenna systems of portable communication devices such as smartphones and laptops (notebooks (NBs)) [1-5]. In particular, not only laptops support wireless LAN (WLAN) but also developments have occurred with respect to 4G/5G wireless systems [6-9]. A system with long-term evolution (LTE) and sub-6G (FR1) applications is thus required. The antennas inside portable devices must cover BW% (bandwidth percentage) = 31.3% for low band; this antenna covered LTE 700 (0.698-0.787 GHz), GSM 850 (0.824-0.894 GHz), and GSM 900 (0.88-0.96 GHz). At the middle band, the bandwidth percentage was 44.5%, and it equalled GSM 1800 (1.71-1.88 GHz)), GSM 1900 (1.85-1.99 GHz), UMTS (1.92-2.17 GHz), and LTE 2500 (2500-2690 MHz). At the high band, BW% was 21% for C band (3.4-4.2 GHz) and BW% = 12.7% for license-assisted access (LAA) (5.15-5.85 GHz) [2-4, 10]. In current commercial communication equipment of STBR (screen-to-body ratio) is significant for user impression; thus, limited space is reserved for antennas. In particular, NBs and tablets have multiple smalldimension antennas above the screen such as two WIFI and

two LTE antennas [6, 11]; it is challenged for small-size antenna design. Thus, the antennas used in wireless equipment are required to be small as well as to exhibit a low profile. Previously executed research [9, 12] has proposed an approach for designing small antennas with a meander architecture. In other research studies [3, 8, 10, 13], meander structures with loop-type antennas were used to achieve a small equipment size. In addition, a bent arm was used by other researchers to design an antenna for multiband applications [12, 14]. The present study was executed to design an antenna exhibiting low as well as high operating bands achieved by applying direct-fed and coupled-fed mechanisms; the derived bands can encompass the LTE, WWAN, C band (n77 and n78), and LAA application bands. In the proposed antenna, the arms have a small size and planar structure.

2. Antenna Design

An FR4 substrate constituted the basis for constructing the proposed antenna in Figures 1(a) and 1(b) and optimized dimensions in Table 1. Fabricated prototype antenna is shown in Figure 2 for measurement; the substrate's dielectric constant, thickness, and loss tangent were determined to be 4.4, 0.4 mm, and 0.02, respectively. An NB and tablet's



(b)

FIGURE 1: (a) Schematic of developed antenna's geometry; (b) definition of component.

Parameter	L	L_1	L_2	L_3	L_4	L_5	L_6	W	W_1	W_2	W_3	W_4	W_5	G
Value (mm)	65	23	8	14.5	21.75	14.25	15	13	3	3	3	4	7	30

ground plane were simulated by a 13 in screen measuring $200 \times 260 \text{ mm}^2$. For the developed antenna, the space it would occupy was determined to measure $65 \times 13 \text{ mm}^2$. Accordingly, the antenna can suitably be placed on the top

side of the compartment of an NB or tablet. The aforementioned direct-fed arm implemented can excite a 1/4 λ fundamental resonant mode at 0.85 GHz as well as upper 3/4 λ and 5.4 λ resonant modes at 2.69 and 4.0 GHz, respectively,





(b)

FIGURE 2: Prototype antenna.

on the right side. In addition, it can excite a $1/4 \lambda$ resonant mode at 5.5 GHz on the left side. Moreover, the coupled-fed arm could excite a fundamental resonant mode at 0.76 GHz and upper resonant modes of 3/4 and 5.4 λ at 2.34 and 3.4 GHz, respectively. Through the integration of the two wide modes at 0.76 and 0.85 GHz, an operating range of 0.698-0.96 GHz (bandwidth of 0.262 GHz or 31.7%) can be obtained. This range covers the GSM 850 and 900 as well as the LTE 700 application bands. The higher modes at 0.76 and 0.85 GHz achieve an operating range of 1.7-2.7 GHz (bandwidth of 0.99 GHz or 45.1%) for the DCS 1800, PCS 1900, UMTS 2100, and LTE 2500 bands as well as the C band range of 3.3-4.2 GHz. The direct-fed arm on the left side has a fundamental resonant mode of 1/4 λ at 5.5 GHz for the LAA (5.15-5.85 GHz). For antenna testing in this study, one side of a coaxial cable was connected to the direct-fed arm and the other side was connected to the contact system ground plane, which was used to provide the RF signal input.

3. Experimental Results and Parametric Study

The derived S_{11} values from the simulation executed through ANSYS HFSS software, and measurement in this study is displayed in Figure 3, indicating good agreement. The testing frequencies obtained for an S_{11} value of -6 dB covered the 2G, 3G, 4G, and 5G systems. Figure 4(a) displays the simulated S_{11} values for different coupled-fed arm lengths. Shortening 10 mm and 15 mm of the couple-fed arm induced the fundamental mode at 0.76 GHz move to high levels of frequency as well as 0.76 GHz higher modes of 2.34 and 3.4 GHz move to a higher frequency. Figures 4(b)–4(d) display the coupled-fed

arm current distribution at 0.76, 2.34, and 3.4 GHz, respectively. For the mode at 0.76 GHz, one null was observed for the couple-fed arm, as shown in Figure 3(b). Thus, the 0.76 GHz fundamental mode was $1/4 \lambda$. For the mode at 2.34 GHz, two null values were observed for the couple-fed arm, as shown in Figure 4(c). Thus, 2.34 GHz is a higher mode of $3/4 \lambda$. For the mode at 3.4 GHz, three null values were observed for the couple-fed arm, as shown in Figure 4(d). Thus, 3.4 GHz is a higher mode of $5/4 \lambda$.

Figure 5(a) displays the simulated S₁₁ values derived when different lengths were considered for the right-side direct-fed arm implemented on the right side. This direct-fed arm had a resonant fundamental mode at 0.85 GHz. When this arm was shortened 5 mm and 8 mm, a 0.85 GHz shift and mismatch were caused, and 0.85 GHz higher mode of 2.69 and 4.0 GHz moved to higher frequencies. Figures 5(b)-5(d) illustrate the simulated current distribution of the rightside direct-fed arm at 0.85, 2.69, and 4.0 GHz, respectively. For the mode at 0.85 GHz, one null was observed for the direct-fed arm, as shown in Figure 5(b). Thus, 0.76 GHz fundamental mode was $1/4 \lambda$. For the mode at 2.69 GHz, two null values were observed for the direct-fed arm, as shown in Figure 5(c). Thus, 2.69 GHz is a higher mode of $3/4 \lambda$. For the mode at 4.0 GHz, three null values were observed for the direct-fed arm, as shown in Figure 5(d). Thus, 4.0 GHz is a higher mode of $5/4 \lambda$.

Figure 6(a) displays the simulated S_{11} values when the left-side direct-fed arm was adjusted in terms of length. The fundamental mode of this arm was 5.5 GHz. 5.5 GHz fundamental mode moved toward higher frequencies as the arm was shortened. Figure 6(b) displays the simulated current



FIGURE 3: S₁₁ values derived from simulation and measurement.



FIGURE 4: (a) S_{11} values derived from simulation for various lengths of the coupled-fed arm of the developed antenna; couple-fed arm's current distribution at (b) 0.76, (c) 2.34, and (d) 3.4 GHz.

distribution of the left-side direct-fed arm at 5.5 GHz. 5.5 GHz fundamental mode was $1/4 \lambda$.

Figure 7(a) displays the simulated S_{11} values obtained with and without a ground strip (matching element). The ground strip influenced antenna matching. Without a ground strip, the low and high modes changed with impedance. With a ground strip, the low operating band ended at 0.96 GHz and the high operating band ended at 5.85 GHz. Figure 7(b) presents the simulated S_{11} values obtained when an inverted L-shaped matching element was and was not used. This element influenced the direct-fed arm at 4 GHz. Without the aforementioned matching element, a significant shift to high frequency occurred. With the matching, the operating band covered the frequency range from 3.3 to 4.2 GHz.

Figure 8 displays the 2D antenna radiation patterns in the *x-z*, *x-y*, and *y-z* planes at 760 MHz, 850 MHz, 2340 MHz, 2690 MHz, 3400 MHz, 4000 MHz, and 5500 MHz. The *x-y* plane pattern is shown to be omnidirectional because 760, 850, and 5500 MHz represent the base modes. Figures 9–11 display the measured efficiency and gain in the lower band as well as in the upper band. The antenna gain was determined to be 0.43-1.94 dBi (1.51 dBi variation) in the lower band (Figure 9), with the



FIGURE 5: (a) S_{11} values derived from the simulation considering different lengths of the right-side direct-fed arm of the developed antenna; current distribution of the direct-fed dipole at (b) 0.85, (c) 2.69, and (d) 4 GHz.



FIGURE 6: (a) S_{11} values derived from simulation considering various lengths of the left-side direct-fed arm for the developed antenna; (b) left-side arm's current distribution at 5.5 GHz.

corresponding radiation efficiency exceeding 55%. The gain was noted to be 0.83–5.67 dBi in the middle band (at 1.71–2.69 GHz; Figure 10), with the corresponding efficiency being 55–83.6%. The gain was determined to be above 3 dBi and 0.9 dBi at 3.2–4.2 and 5.15–5.85 GHz, respectively, in Figure 11 with the corresponding efficiency ranging from 60% to 86.4%. Measured resonant, gain, efficiency, and bandwidth are arranged in Table 2. The proposed antenna has suitable and stable radiation efficiency.

Performance was compared between present antennas with reference to some portable device antenna. Table 3 displays dimension, volume, bandwidth, S_{11} , and RLC element. In [1, 12, 14], LC element is used and bent-type structure greatly reduces dimension, while the volume is larger, and cover bandwidth is less than present antenna. In [4, 5, 9, 10], LC element is not used, and antenna occupies a smaller area, but volume and cover bandwidth are less than present antenna. Furthermore, present antenna can cover



FIGURE 7: (a) S_{11} values derived from simulation of developed antenna without and with a ground strip; (b) S_{11} values derived from simulation of developed antenna without and with an inverted L-shaped matching element.



(c) FIGURE 8: Continued.



FIGURE 8: Measured 2D antenna radiation patterns at *x-y*, *x-z*, and *y-z* planes: (a) 760 MHz; (b) 850 MHz; (c) 2340 MHz; (d) 2690 MHz; (e) 3400 MHz; (f) 4000 MHz; (g) 5500 MHz.



FIGURE 9: Gain and radiation efficiency measured for the developed antenna at 700-960 MHz.



FIGURE 10: Gain and radiation efficiency measured for the developed antenna at 1710-2690 MHz.



FIGURE 11: Gain and radiation efficiency measured for the developed antenna at 3300-5850 MHz.

ABLE 2: Arrangement of measure	d resonant, gain, efficienc	y, and bandwidth of	proposed antenna.
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Resonance frequency (GHz)	Variation gain (dBi)	Variation efficiency (%)	Variation BW (GHz)	Resonance frequency (GHz)	Variation gain (dBi)
0.76 and 0.85	1.51 (0.43-1.94)	0.3 (55.8-56.1)	0.262 (0.698-0.96)	0.76 and 0.85	1.51 (0.43-1.94)
2.34 and 2.69	4.84 (0.83-5.67)	28.6 (55.0-83.6)	0.98 (1.71-2.69)	2.34 and 2.69	4.84 (0.83-5.67)
3.4 and 4.0	0.97 (4.45-5.42)	18.4 (68.0-86.4)	0.9 (3.3-4.2)	3.4 and 4.0	0.97 (4.45-5.42)

TABLE 3: Comparison between the present antenna and reference antennas.

Reference	Area (mm)	Volume (mm ³)	Bandwidth (MHz)	S_{11} (dB)	LC element
[1]	$15 \times 28 \times 4$	1680	804-967/1665-3280	-6	Yes
[2]	$60 \times 15 \times 2.8$	2520	698-1100/1640-2830	-6	No
[4]	$55 \times 12 \times 5$	3300	794-962/1710-2730	-6	No
[10]	$32 \times 18 \times 3$	1728	824-960/1710-2170/	-6	No
[12]	$60 \times 8 \times 5$	2400	800-960/1710-2580	-6	Yes
[11]	$75 \times 12 \times 4$	3600	798-787/2305-2400/2500-2690/3400-3600	-6	No
[14]	$30 \times 10 \times 3$	900	698-960/1710-2170/3400~3800	-6	Yes
[5]	$28 \times 15 \times 4$	1500	824-960/1710-2170/2300-2690	-6	No
[9]	$60 \times 7 \times 3$	1260	698-960/1710-2170/2300-2690	-6	No
Presents antenna	$65 \times 13 \times 0.4$	845	698-960/1710-2170/2300-2690/3300-4200/5150-5850	-6	No

bandwidth (698–960 MHz, 1710–2690 MHz, 3300–4200 MHz, and 5150–5850 MHz) as well as smaller size and planer structure than other reference antennas.

4. Conclusions

In this paper, a low profile, planar, and small-size antenna design is proposed for 5G portable communication equipment. The proposed antenna uses seven resonant modes to cover the frequency ranges of 0.69–0.96, 1.71–2.69, 3.2–4.3, and 5.15–5.85 GHz. The operating range of the antenna covers the LTE 700/2500, GSM, UMTS, C band (n77 and n78), and license-assisted access (LAA). The proposed antenna has small dimensions, two wide bands of operation, omnidirectional radiation in low-frequency bands, and reasonable antenna radiation efficiency and gain. The measured results of the proposed antenna signify its suitability for portable devices.

Data Availability

The data used to support the findings of this study are available from the corresponding author on request.

Conflicts of Interest

The authors declare that they have no conflicts of interest.

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